

Use of own programmable calculator is allowed.

Problem 1 (max 6 points)

Transfer function matrix of a grid-connected battery-interfacing inverter is given at open-loop in (1) and at closed-loop in (2). What transfer function should be analyzed in the following cases described below.

- The discharging current of the battery i_{bat} is controlled using DC-current feedback and the controller transfer function should be designed. DC-side dynamics are assumed to be dominated by d-components (real power).
- Current control is implemented in dq-domain. *Crossover frequency* and *phase margin* of the AC-current control should be evaluated. It is assumed that PI-controlled parameters are already selected, i.e., controller transfer function is known.
- Control design is already finished and the effect of grid voltage harmonics on the generated output currents should be analyzed.
- Control bandwidth* of the AC-current control should be determined.
- Output admittance d-component should be re-shaped to have higher magnitude at the fifth harmonic. Changing the size of passive components is not allowed.

$$\begin{bmatrix} \hat{i}_{\text{bat}} \\ \hat{i}_{\text{od}} \\ \hat{i}_{\text{oq}} \end{bmatrix} = \begin{bmatrix} G_{11} & G_{12} & G_{13} & G_{14} & G_{15} \\ G_{21} & G_{22} & G_{23} & G_{24} & G_{25} \\ G_{31} & G_{32} & G_{33} & G_{34} & G_{35} \end{bmatrix} \begin{bmatrix} \hat{v}_{\text{bat}} \\ \hat{v}_{\text{od}} \\ \hat{v}_{\text{oq}} \\ \hat{d}_d \\ \hat{d}_q \end{bmatrix} \quad (1)$$

$$\begin{bmatrix} \hat{i}_{\text{bat}} \\ \hat{i}_{\text{od}} \\ \hat{i}_{\text{oq}} \end{bmatrix} = \begin{bmatrix} G_{11} & G_{12} & G_{13} & G_{14} & G_{15} \\ G_{21} & G_{22} & G_{23} & G_{24} & G_{25} \\ G_{31} & G_{32} & G_{33} & G_{34} & G_{35} \end{bmatrix} \begin{bmatrix} \hat{v}_{\text{bat}} \\ \hat{v}_{\text{od}} \\ \hat{v}_{\text{oq}} \\ \hat{i}_{\text{od}}^* \\ \hat{i}_{\text{oq}}^* \end{bmatrix} \quad (2)$$

Problem 2 (max 6 points)

Control block diagram of a phase-locked-loop is as shown in Figure 1. The feedforward term ω_{ff} is a constant which improves start-up. The Park's transformation can be linearized as $\hat{v}'_q = \hat{v}_q - V_d \hat{\theta}$ where \hat{v}_q denotes the ideal grid voltage q-component.

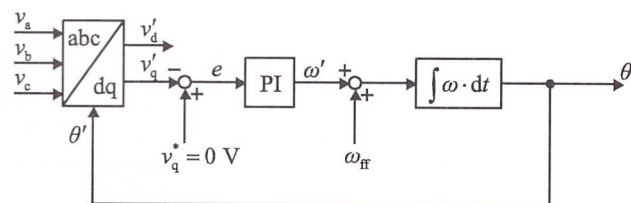


Figure 1: Phase-locked-loop.

- Draw the linearized control block diagram and define control loop gain of the PLL.
- Solve transfer function from the reference input \hat{v}_q^* to the controlled variable \hat{v}'_q in dq-domain.
- The transfer function from reference to the controlled variable can be written as a second-order system as in (3). Find out the damping ratio ζ and natural frequency ω_n in terms of controller parameters. You can assume that the controller transfer function is as given in (4).

$$G = (2\zeta\omega_n s + \omega_n^2) / (s^2 + 2\zeta\omega_n s + \omega_n^2) \quad (3)$$

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$$G_c = \frac{(-1) \cdot K (s/\omega_z + 1)}{s} \quad (4)$$

Problem 3 (max 6 points)

Figure 2 shows a simple power electronics -based power system. Answer the following questions:

- Which converter is responsible of frequency control of the AC line? Justify how you determine this from the figure.
- What variables are controlled using cascaded control scheme?
- Which converter is responsible of regulating the RMS value of AC voltage? Justify based on Figure 2.
- How can you decide the minimum voltage of the DC voltage source? (short answer please, no equations needed)
- Consider you need to define an open-loop small-signal model for the load converter. What are the output variables of the small-signal model? Justify their selection.

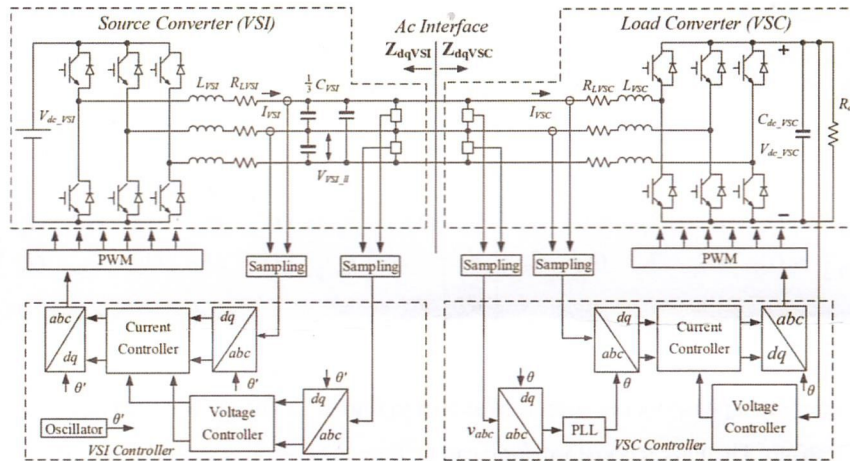


Figure 2: Control system of a power-electronics based AC system.

Problem 4 (max 6 points)

Three-phase LCL-filter is shown in Figure 3. Solve the average state-space model of the filter in the dq-domain. You can assume that the three-phase input and output voltages are balanced. Draw the electrical circuits of the filter in dq-domain (separately for d and q-components).

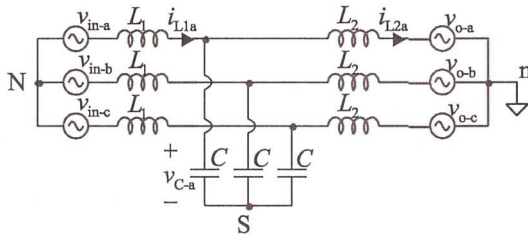


Figure 3: Three-phase LCL-type filter.

Hints:

$$\frac{d}{dt} i_{L1}^{dq} = f(v_{NS}, v_{in-d}, v_{in-q}, v_{C-d}, v_{C-q})?$$

$$\frac{d}{dt} v_C^{dq} = f(i_{L1d}, i_{L1q}, i_{L2d}, i_{L2q})?$$

$$\frac{d}{dt} i_{L2}^{dq} = f(v_{NS}, v_{C-d}, v_{C-q}, v_{o-d}, v_{o-q})?$$

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Problem 5 (max 6 points)

Grid voltage in unbalanced condition is given in stationary frame according to (5).

- How is the grid voltage seen in the dq-domain? Give the amplitude and frequencies of different components.
- A three-phase inverter is connected to the unbalanced grid defined in (5). The fundamental frequency of the grid is 50 Hz. A conventional SRF-PLL is used according Figure 1. How much does the PLL loop gain attenuate the effect of unbalance when it is as shown in Figure 4?
- How does the unbalance affect power quality of the inverter?
- What can you do to mitigate the effect of unbalance even more?

$$\mathbf{v}_{\text{grid}}^{\alpha\beta} = 230 \cdot e^{j\omega_s t} + 100 \cdot e^{j(-\omega_s)t} \quad (5)$$

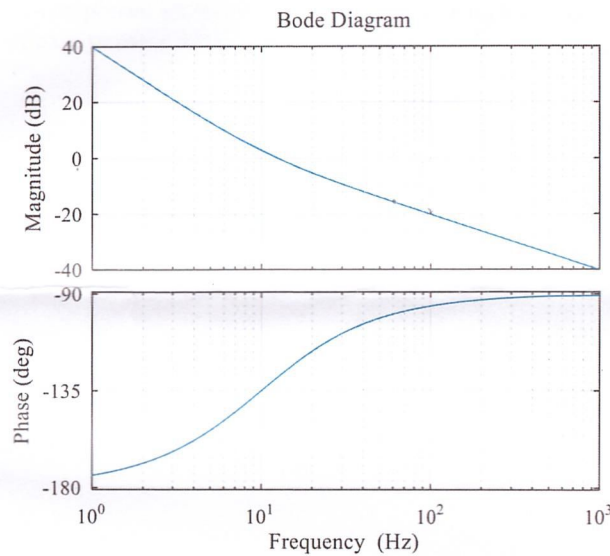


Figure 4: Control loop gain of the PLL in Problem 5.

Equations...

$$\mathbf{x}^{\alpha\beta} = \mathbf{x}^{\text{dq}} \cdot e^{j\omega_s t}$$

$$e^{-jx} = \cos(x) - j\sin(x)$$

$$\mathbf{T}^{\text{abc} \rightarrow \alpha\beta} \cdot [k \ k \ k]^T = [0 \ 0 \ k]^T$$